



WORLD INTELLECTUAL PROPERTY ORGANIZATION
International Bureau

(51) International Patent Classification ⁶ : H04B 1/40		A1	(11) International Publication Number: WO 99/45654
			(43) International Publication Date: 10 September 1999 (10.09.99)
(21) International Application Number: PCT/US99/04545		(81) Designated States: AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, CA, CH, CN, CU, CZ, DE, DK, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MD, MG, MK, MN, MW, MX, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, UA, UG, UZ, VN, YU, ZW, ARIPO patent (GH, GM, KE, LS, MW, SD, SL, SZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GW, ML, MR, NE, SN, TD, TG).	
(22) International Filing Date: 2 March 1999 (02.03.99)			
(30) Priority Data: 60/076,520 2 March 1998 (02.03.98) US 09/259,242 26 February 1999 (26.02.99) US			
(71) Applicant: QUALCOMM INCORPORATED [US/US]; 6455 Lusk Boulevard, San Diego, CA 92121 (US).			
(72) Inventors: CICCARELLI, Steven, C.; 45272 Callesita Ordenes, Temecula, CA 92592 (US). YOUNIS, Saed, G.; 12767 Jordan Ridge Court, San Diego, CA 92130 (US).			
(74) Agents: MILLER, Russell, B. et al.; Qualcomm [®] Incorporated, 6455 Lusk Boulevard, San Diego, CA 92121 (US).		<p>Published</p> <p><i>With international search report.</i></p> <p><i>Before the expiration of the time limit for amending the claims and to be republished in the event of the receipt of amendments.</i></p>	

(57) Abstract

signals from the first frequency range to the common intermediate frequency range. A second downconverter (24) is selectively coupled by a switch (16) to either an output of the second band selection filter (12) or an output of the third band selection filter (14), and downconverts signals from either the second frequency range or the third frequency range to the common intermediate frequency range. The second downconverter has an input coupled to a frequency doubling circuit (26). Switching circuitry (28) selectively couples one of either a first oscillating signal (30) from a voltage controlled oscillator (VCO) (34) having a VCO frequency range or a second oscillating signal (32) at a second frequency that is outside the VCO frequency range to an input of the first downconverter (20) and an input of the frequency doubling circuit (26).

FOR THE PURPOSES OF INFORMATION ONLY

Codes used to identify States party to the PCT on the front pages of pamphlets publishing international applications under the PCT.

AL	Albania	ES	Spain	LS	Lesotho	SI	Slovenia
AM	Armenia	FI	Finland	LT	Lithuania	SK	Slovakia
AT	Austria	FR	France	LU	Luxembourg	SN	Senegal
AU	Australia	GA	Gabon	LV	Latvia	SZ	Swaziland
AZ	Azerbaijan	GB	United Kingdom	MC	Monaco	TD	Chad
BA	Bosnia and Herzegovina	GE	Georgia	MD	Republic of Moldova	TG	Togo
BB	Barbados	GH	Ghana	MG	Madagascar	TJ	Tajikistan
BE	Belgium	GN	Guinea	MK	The former Yugoslav Republic of Macedonia	TM	Turkmenistan
BF	Burkina Faso	GR	Greece	ML	Mali	TR	Turkey
BG	Bulgaria	HU	Hungary	MN	Mongolia	TT	Trinidad and Tobago
BJ	Benin	IE	Ireland	MR	Mauritania	UA	Ukraine
BR	Brazil	IL	Israel	MW	Malawi	UG	Uganda
BY	Belarus	IS	Iceland	MX	Mexico	US	United States of America
CA	Canada	IT	Italy	NE	Niger	UZ	Uzbekistan
CF	Central African Republic	JP	Japan	NL	Netherlands	VN	Viet Nam
CG	Congo	KE	Kenya	NO	Norway	YU	Yugoslavia
CH	Switzerland	KG	Kyrgyzstan	NZ	New Zealand	ZW	Zimbabwe
CI	Côte d'Ivoire	KP	Democratic People's Republic of Korea	PL	Poland		
CM	Cameroon	KR	Republic of Korea	PT	Portugal		
CN	China	KZ	Kazakhstan	RO	Romania		
CU	Cuba	LC	Saint Lucia	RU	Russian Federation		
CZ	Czech Republic	LI	Liechtenstein	SD	Sudan		
DE	Germany	LK	Sri Lanka	SE	Sweden		
DK	Denmark	LR	Liberia	SG	Singapore		
EE	Estonia						

**METHOD AND APPARATUS FOR DOWNCONVERTING
SIGNALS TRANSMITTED USING A PLURALITY OF
MODULATION FORMATS TO A COMMON INTERMEDIATE
FREQUENCY RANGE**

5

BACKGROUND OF THE INVENTION

10 I. **Field of the Invention**

The present invention relates generally to mobile radio communication systems. More particularly, the present invention relates to mobile radio units that can receive and demodulate signals at different frequency bands that have been modulated in accordance with multiple modulation formats. Even more particularly, the present invention relates to a novel and improved receiver circuit that downconverts such signals at different frequency bands to a common frequency band using a minimal number of hardware components.

20 II. **Description of the Prior Art:**

Currently, mobile radio telephones typically support communications in a number of different modes corresponding to different modulation formats. For example, there are so-called "dual-mode" mobile radio telephones that support communications using both analog and code division multiple access (CMDA) signals. As the number of modes that the phone supports increases, the circuitry required in the phone for downconverting and sampling the input signals typically increases in complexity. This occurs because, in multi-mode phones, the input signals are received at different frequency bands depending on the operating mode, and downconverting and sampling the signals from each of the frequency bands typically requires separate circuitry for each band. It would therefore be desirable to have a receiver design that could be used in multi-mode phones for downconverting and sampling the input signals which used common hardware for the downconversion and sampling operations, thereby minimizing the hardware components required for operation of a multi-mode phone.

SUMMARY OF THE INVENTION

There present invention is directed to a receiver that downconverts input signals modulated using first, second, third and fourth modulation formats to a common intermediate frequency range. The first and second modulation formats are transmitted to the receiver in a first frequency range, the third modulation format is transmitted to the receiver in a second frequency range, and the fourth modulation format is transmitted to the receiver in a third frequency range. The input signals are provided to first, second and third band selection filters that respectively select first, second and third frequency ranges. A first downconverter is coupled to an output of the first band selection filter, and downconverts signals from the first frequency range to the common intermediate frequency range. A second downconverter is selectively coupled by a switch to either an output of the second band selection filter or an output of the third band selection filter, and downconverts signals from either the second frequency range or the third frequency range to the common intermediate frequency range. The second downconverter has an input coupled to a frequency doubling circuit. Switching circuitry selectively couples one of either a first oscillating signal from a voltage controlled oscillator (VCO) having a VCO frequency range or a second oscillating signal at a second frequency that is outside the VCO frequency range to an input of the first downconverter and an input of the frequency doubling circuit.

In a preferred embodiment, the VCO is responsive to control circuitry and selectively generates the first oscillating signal at a channel frequency associated with the first modulation format, and the first downconverter mixes the output of the first band selection filter with the first oscillating signal at the channel frequency associated with the first modulation format in order to downconvert signals modulated in accordance with the first modulation format from the first frequency range to the common intermediate frequency range. The VCO further selectively generates the first oscillating signal at a channel frequency associated with the second modulation format, and the first downconverter mixes the output of the first band selection filter with the first oscillating signal at the channel frequency associated with the second modulation format in order to downconvert signals modulated in accordance with the second modulation format from the first frequency range

to the common intermediate frequency range. In addition, in the preferred embodiment, the VCO further selectively generates the first oscillating signal at a channel frequency associated with the third modulation format, and the first oscillating signal at the channel frequency associated with the third modulation format is provided to the frequency doubling circuit. The second downconverter then mixes the output of the second band selection filter with an output of the frequency doubling circuit when the first oscillating signal at a channel frequency associated with the third modulation format is provided to the frequency doubling circuit in order to downconvert signals modulated in accordance with the third modulation format from the second frequency range to the common intermediate frequency range.

The receiver of the present invention also preferably includes a mixer that forms the second oscillating signal at the second frequency by offsetting the frequency of the first oscillating signal. The second oscillating signal is selectively provided to the frequency doubling circuit, and the second downconverter mixes the output of the third band selection filter with the output of the frequency doubling circuit when the second oscillating signal is provided to the frequency doubling circuit in order to downconvert signals modulated in accordance with the fourth modulation format from the third frequency range to the common intermediate frequency range.

In accordance with a further aspect, the receiver of the present invention includes one or more channel selection filters that are coupled to outputs of the first and second downconverters. The channel selection filter(s) function to filter the downconverted signals output by the first and second downconverters. A sampling circuit is coupled to an output of the channel selection filter(s). The sampling circuit selectively samples the downconverted signals with a sampling clock that alternatively samples signals at either a first or second sampling rate. The sampling clock is coupled to a third oscillating signal at a third frequency, and the third oscillating signal is an input to the mixer that forms the second oscillating signal at the second frequency. The first sampling rate provided by the sampling clock is equal to the third frequency divided by x , and the second sampling rate provided by the sampling clock is equal to the third frequency divided by y , wherein x and y are integers and, in one example, are equal to 3 and 15. This aspect invention minimizes excess hardware in the receiver, since the third oscillating signal at the third frequency

is used both in the frequency downconversion process as well as in the sampling process. In one embodiment, the sampling circuit samples downconverted signals modulated in accordance with either the second, third or fourth modulation formats in accordance with the first sampling rate, and
5 the sampling circuit samples downconverted signals modulated in accordance with the first modulation format in accordance with the second sampling rate.

In one embodiment, only a single channel selection filter is coupled to outputs of the first and second downconverters, and the single channel selection filter operates to filter downconverted signals modulated in
10 accordance with the first and second modulation formats output by the first downconverter, and the single channel selection filter further filters downconverted signals modulated in accordance with the third and fourth modulation formats output by the second downconverter.

In a further embodiment, first and second channel selection filters
15 are coupled to outputs of the first and second downconverters. The first channel selection filter operates to filter downconverted signals modulated in accordance with the first modulation format output by the first downconverter, the second channel selection filter operates to filter downconverted signals modulated in accordance with the second modulation format output by the first
20 downconverter, the second channel filter also filters downconverted signals modulated in accordance with the third modulation format output by the second downconverter, and the second channel selection filter further filters downconverted signals modulated in accordance with the fourth modulation format output by the second downconverter.

In a still further embodiment, first, second and third channel selection filters are coupled to outputs of the first and second downconverters. The first channel selection filter operates to filter downconverted signals modulated in accordance with the first modulation format output by the first downconverter, the second channel selection filter operates to filter
30 downconverted signals modulated in accordance with the second modulation format output by the first downconverter, the second channel filter also filters downconverted signals modulated in accordance with the third modulation format output by the second downconverter, and the third channel selection filter operates to filter downconverted signals modulated in accordance with
35 the fourth modulation format output by the second downconverter.

In the above embodiments, two downconverters were coupled to outputs of the band selection filters. In a further embodiment, a third downconverter is coupled to an output of the first band selection filter, and also functions to downconvert signals from the first frequency range to the common intermediate frequency range. In this embodiment, the switching circuitry selectively couples one of either the first oscillating signal from the VCO or the second oscillating signal to inputs of the first and third downconverters and an input of the frequency doubling circuit. Whereas in the previous embodiments input signals modulated in accordance with the second modulation format were downconverted using the first downconverter, in this embodiment the third downconverter mixes the output of the first band selection filter with the first oscillating signal at the channel frequency associated with the second modulation format in order to downconvert signals modulated in accordance with the second modulation format from the first frequency range to the common intermediate frequency range. In this embodiment, one, two or three channel selection filters is/are alternatively coupled to outputs of the first, second and third downconverters, and the channel selection filter(s) operate to filter the downconverted signals modulated in accordance with the first, second, third and fourth modulation formats.

BRIEF DESCRIPTION OF THE DRAWINGS

The features, objects and advantages of the present invention will become more apparent from the detailed description set forth below when taken in conjunction with the drawings in which like reference characters identify corresponding elements and wherein:

FIG. 1 is a block diagram of a receiver that downconverts signals transmitted at three different frequency ranges and modulated using first, second, third and fourth modulation formats to a common intermediate frequency range using three downconverters, in accordance with a preferred embodiment of the present invention. In the receiver of FIG. 1, three channel selection filters are used to process the downconverted signals output by the three downconverters.

FIG. 2 is a block diagram of a receiver that downconverts signals transmitted at three different frequency ranges and modulated using first,

second, third and fourth modulation formats to a common intermediate frequency range using three downconverters, in accordance with an alternative preferred embodiment of the present invention. In the receiver of FIG. 2, two channel selection filters are used to process the downconverted signals output
5 by the three downconverters.

FIG. 3 is a block diagram of a receiver that downconverts signals transmitted at three different frequency ranges and modulated using first, second, third and fourth modulation formats to a common intermediate frequency range using three downconverters, in accordance with a still further preferred embodiment of the present invention. In the receiver of FIG. 3, a
10 single channel selection filter is used to process the downconverted signals output by the three downconverters.

FIG. 4 is a block diagram of a receiver that downconverts signals transmitted at three different frequency ranges and modulated using first, second, third and fourth modulation formats to a common intermediate frequency range using two downconverters, in accordance with a preferred
15 embodiment of the present invention. In the receiver of FIG. 4, three channel selection filters are used to process the downconverted signals output by the two downconverters.

FIG. 5 is a block diagram of a receiver that downconverts signals transmitted at three different frequency ranges and modulated using first, second, third and fourth modulation formats to a common intermediate frequency range using two downconverters, in accordance with a further alternative preferred embodiment of the present invention. In the receiver of
20 FIG. 5, two channel selection filters are used to process the downconverted signals output by the two downconverters.

FIG. 6 is a block diagram of a receiver that downconverts signals transmitted at three different frequency ranges and modulated using first, second, third and fourth modulation formats to a common intermediate frequency range using two downconverters, in accordance with a still further preferred embodiment of the present invention. In the receiver of FIG. 6, a
30 single channel selection filter is used to process the downconverted signals output by the two downconverters.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

Referring now to FIG. 1, there is shown a block diagram of a receiver 2 for use in a multi-mode mobile radio telephone that alternatively supports communications using any one of four different modulation formats. The receiver downconverts input signals modulated using one of first, second, third and fourth modulation formats to a common intermediate frequency range. In the embodiment shown, the first, second, third and fourth modulation formats correspond respectively to analog modulation, CDMA modulation, PCS modulation and GPS modulation, although it will be understood by those skilled in the art that the architecture of the present invention could be applied for downconverting signals modulated in accordance with other formats. The first and second modulation formats are transmitted to the receiver in a first frequency range, the third modulation format is transmitted to the receiver in a second frequency range, and the fourth modulation format is transmitted to the receiver in a third frequency range. In the embodiment shown, the first and second frequency ranges correspond to the frequency bands typically used for analog and CDMA cellular telephone transmissions, i.e., 869-894 MHz; the third frequency range corresponds to the frequency band typically used for PCS transmissions, i.e., 1930-1990 MHz; and the fourth frequency range corresponds to the frequency band typically used for GPS transmissions, i.e., a 2 MHz band centered at 1575.42 MHz.

Input signals received by the receiver are provided to a first band selection filter 10 that selects the first frequency range (e.g., the frequency range corresponding to the analog and CDMA bands), a second band selection filter 12 selects the second frequency range (e.g., the PCS band), and a third band selection filter 14 selects the third frequency range (e.g., the GPS band).

A downconverter (or mixer) 20 is coupled to an output of the first band selection filter 10, and downconverts signals modulated in accordance with the first modulation format (e.g., analog signals) output by the first band selection filter 10 from the first frequency range to the common intermediate frequency range. The common intermediate frequency range is centered at an IF center frequency which, in the preferred embodiment, is about 183.48 MHz. The downconverter 20 translates the output of the first band selection filter 10

(i.e., the entire analog frequency band) down the frequency spectrum to the intermediate frequency range such that the channel of interest (i.e., the channel in the analog frequency band to which the receiver is being tuned) is centered at the IF center frequency.

5 A second downconverter (or mixer) 22 is also coupled to an output of the first band selection filter 10, and downconverts signals modulated in accordance with the second modulation format (e.g., CDMA signals) output by the first band selection filter 10 from the first frequency range to the common intermediate frequency range. The downconverter 22 translates the output of
10 the first band selection filter 10 (i.e., the entire CDMA frequency band) down the frequency spectrum to the intermediate frequency range such that the channel of interest (i.e., the channel in the CDMA frequency band to which the receiver is being tuned) is centered at the IF center frequency.

 A third downconverter (or mixer) 24 is selectively coupled by a
15 switch 16 to either an output of the second band selection filter 12 or an output of the third band selection filter 14. The third downconverter 24 functions to downconvert either (i) signals modulated in accordance with the third modulation format (e.g., PCS signals) output by the second band selection filter 12 from the second frequency range to the common intermediate frequency
20 range, or (ii) signals modulated in accordance with the fourth modulation format (e.g., GPS signals) output by the third band selection filter 14 from the third frequency range to the common intermediate frequency range. Depending on the position of switch 16, the downconverter 24 either translates the output of the second band selection filter 12 (i.e., the entire PCS frequency
25 band) down the frequency spectrum to the intermediate frequency range such that the channel of interest (i.e., the channel in the PCS frequency band to which the receiver is being tuned) is centered at the IF center frequency, or the downconverter 24 translates the output of the third band selection filter 14 (i.e., the entire GPS frequency band) down the frequency spectrum to the
30 intermediate frequency range such that the channel of interest (i.e., the channel in the GPS frequency band to which the receiver is being tuned) is centered at the IF center frequency. When switch 16 couples the output of the second band selection filter 12 to the third downconverter 24, switch 29 couples the output of the third downconverter 24 to the second channel selection filter 62 (described
35 below). Similarly, when switch 16 couples the output of the third band

selection filter 14 to the third downconverter 24, switch 29 couples the output of the third downconverter 24 to the third channel selection filter 64 (also described below).

The third downconverter 24 has an input coupled to a frequency
5 doubling circuit 26. Switching circuitry 28 selectively provides one of either a first oscillating signal 30 from a voltage controlled oscillator (VCO) 34 having a VCO frequency range or a second oscillating signal 32 at a second frequency that is outside the VCO frequency range to inputs of the first and second downconverters 20, 22 and to an input of the frequency doubling circuit 26. In
10 the preferred embodiment, the frequency range of VCO 34 is 1052-1087 MHz.

The VCO 34 is responsive to control circuitry 36 that causes the VCO 34 to generate the first oscillating signal 30 at a frequency associated with an analog channel to which the receiver is being tuned when analog signals are being received at antenna 40 (i.e., when the mobile radio is operating in the
15 analog mode.) When such signals are being received, the first downconverter 20 mixes the output of the first band selection filter 10 with the first oscillating signal 30 at the frequency associated with the analog channel to which the receiver is being tuned in order to downconvert analog signals from the first frequency range to the common intermediate frequency range. In the
20 downconverted signal, the channel of interest (i.e., the analog channel to which the receiver is being tuned) is centered at the IF center frequency.

The control circuitry 36 further causes the VCO 34 to generate the first oscillating signal 30 at a frequency associated with a CDMA channel to which the receiver is being tuned when CDMA signals are being received at
25 antenna 40 (i.e., when the mobile radio unit is operating in the CDMA mode.) When such signals are being received, the second downconverter 22 mixes the output of the first band selection filter 10 with the first oscillating signal 30 at the frequency associated with the CDMA channel to which the receiver is being tuned in order to downconvert analog signals from the first frequency range to
30 the common intermediate frequency range. In the downconverted signal, the channel of interest (i.e., the CDMA channel to which the receiver is being tuned) is centered at the IF center frequency.

The control circuitry 36 further causes the VCO 34 to generate the first oscillating signal 30 at a frequency associated with a PCS channel to which
35 the receiver is being tuned when PCS signals are being received at antenna 40

(i.e., when the mobile radio unit is operating in the PCS mode.) When such signals are being received, the first oscillating signal 30 at a frequency associated with a PCS channel to which the receiver is being tuned is provided to the frequency doubling circuit 26, and the third downconverter 24 mixes the
5 output of the second band selection filter 14 with an output of the frequency doubling circuit 26 in order to downconvert PCS signals from the second frequency range to the common intermediate frequency range. In the downconverted signal, the channel of interest (i.e., the PCS channel to which the receiver is being tuned) is centered at the IF center frequency.

10 In a preferred embodiment, when the receiver is operating in either the analog or CDMA modes, the first oscillating signal 30 provided by the VCO 34 is within a frequency range of 1052.52 – 1077.45 MHz. This frequency range allows the downconverters 20, 22 to downconvert any channel of interest in either the analog or CDMA band to the IF center frequency.
15 Similarly, in the preferred embodiment, when the receiver is operating in the PCS mode, the first oscillating signal 30 provided by the VCO 34 is within a frequency range of 1056.74 – 1086.74 MHz. This frequency range allows the downconverter 24 to downconvert any channel of interest in the PCS band to the IF center frequency.

20 When the mobile radio unit is operating in the GPS mode and GPS signals are received at antenna 42, a mixer 50 forms the second oscillating signal 32 at a frequency that is outside the frequency range of VCO 34 by offsetting the frequency of the first oscillating signal 30. This frequency offsetting may be performed using, for example, an image reject mixer (IRM)
25 50. The frequency of the second oscillating signal 32 is associated with a GPS channel to which the receiver is being tuned. When the mobile radio unit is operating in the GPS mode, the second oscillating signal 32 at a frequency associated with a GPS channel to which the receiver is being tuned is provided to the frequency doubling circuit 26, and the third downconverter 24 mixes the
30 output of the third band selection filter 14 with an output of the frequency doubling circuit 26 in order to downconvert GPS signals from the third frequency range to the common intermediate frequency range. In the downconverted signal, the channel of interest (i.e., the GPS channel to which the receiver is being tuned) is centered at the IF center frequency.

In the preferred embodiment, when the receiver is operating in the GPS mode, the second oscillating signal 32 provided to doubling circuit 26 is within a frequency range that is centered about 879.45 MHz. This frequency range allows the downconverter 24 to downconvert any channel of interest in the GPS band to the IF center frequency.

As mentioned above, the mixer 50 forms the second oscillating signal 32 at a frequency that is outside the frequency range of VCO 34 by offsetting the frequency of the first oscillating signal 30. In the preferred embodiment, this frequency offsetting is performed by mixing the output of the VCO 34 with a signal 33 generated by oscillator 35. Signal 33 has a frequency that is equal to three times a sampling frequency (F_s). The sampling frequency (F_s) associated with oscillator 35 is preferably selected such that the sampling frequency and the center IF frequency (f_{IF}) discussed above are related in accordance with equation (1) below:

$$f_{IF} = [(2k + 1)/4] * (F_s), \text{ where } k = 0, 1, 2, \dots \quad (1)$$

The receiver of the FIG. 1 also preferably includes a plurality of channel selection filters 60, 62, 64 for selecting or tuning to specified channels associated with each of the four modulation formats. The channel selection filters 60, 62, 64 are coupled to outputs of the first, second and third downconverters, and filter the downconverted signals output by the first, second and third downconverters. A sampling circuit 70 is coupled to an output of the channel selection filters by anti-alias filters 80, 82. The sampling circuit 70 selectively samples the downconverted signals with a sampling clock 72 that alternatively samples signals at either a first or second sampling rate. The sampling clock 72 is coupled to signal 33 which, as mentioned above, is also provided as an input to the mixer 50 that forms the second oscillating signal 32. The first sampling rate provided by the sampling clock is preferably equal to $3 * F_s$ (i.e., the frequency of signal 33) divided by 3, and the second sampling rate provided by the sampling clock is preferably equal to $(3 * F_s)$ divided by 15. These two sampling rates allow the receiver of the present invention to sample either analog, CDMA, PCS or GPS signals using a common sampling circuit 70. In particular, the present invention uses the first sampling rate to sample CDMA, PCS and GPS signals, and the second sampling rate to sample analog signals. This aspect of the invention minimizes excess hardware

in the receiver, since the oscillating signal 33 is used both in the frequency downconversion process as well as in the sampling process.

Also in the preferred embodiment, a radio frequency integrated circuit chip is preferably used to implement all the components positioned
5 within dotted line 5.

Referring now to FIG. 2, there is shown a block diagram of a receiver that downconverts signals modulated using first, second, third and fourth modulation formats to a common intermediate frequency range using three downconverters, in accordance with an alternative preferred embodiment
10 of the present invention. The receiver shown in FIG. 2 is substantially the same as the receiver of FIG. 1, except in the receiver of FIG. 2 only two channel selection filters 60, 62A are used to process the downconverted signals output by the three downconverters. Thus, in the receiver of FIG. 2, the second channel selection filter 62A filters downconverted signals modulated in
15 accordance with the second modulation format (e.g., CDMA signals) output by the second downconverter 22, the second channel filter 62A also filters downconverted signals modulated in accordance with the third modulation format (e.g., PCS signals) output by the third downconverter 24, and the second channel selection filter 62A further filters downconverted signals modulated in
20 accordance with the fourth modulation format (e.g., GPS signals) output by the third downconverter 24.

Referring now to FIG. 3, there is shown a block diagram of a receiver that downconverts signals modulated using first, second, third and fourth modulation formats to a common intermediate frequency range using
25 three downconverters, in accordance with a further alternative preferred embodiment of the present invention. The receiver shown in FIG. 3 is substantially the same as the receiver of FIG. 1, except in the receiver of FIG. 3 only a single channel selection filter 60A is used to process the downconverted signals output by the three downconverters.

FIG. 4 is a block diagram of a receiver that downconverts signals modulated using first, second, third and fourth modulation formats to a common intermediate frequency, in accordance with a preferred embodiment of the present invention. In contrast to the receiver of FIGS. 1-3, in the receiver of FIG. 4 only two downconverters 20, 24 are used to perform the frequency
35 downconversions for all four operating modes. Thus, in FIG. 4, only

downconverter 20 is coupled to an output of the first band selection filter 10. Downconverter 20 downconverts signals modulated in accordance with either the first modulation format (e.g., analog signals) or the second modulation format (e.g., CDMA signals) output by the first band selection filter 10 from the first frequency range to the common intermediate frequency range. In this embodiment, the downconverter 20 translates the output of the first band selection filter 10 (i.e., either the entire analog or CDMA frequency band) down the frequency spectrum to the intermediate frequency range such that the channel of interest (i.e., the channel in the analog or CDMA frequency band to which the receiver is being tuned) is centered at the IF center frequency. The remaining components of the receiver of FIG. 4 function substantially the same as the corresponding components described above in connection with FIG. 1, except in FIG. 4 the first oscillating signal 30 that was previously provided to mixer 22 in FIG. 1 is provided instead to mixer 20 in FIG. 4 when the receiver is operating in the CDMA mode. In addition, in the embodiment of FIG. 4, a further switch 29A is provided for directing the output of mixer 20 to the second channel selection filter 62 when the receiver is operating in the CDMA mode.

Referring now to FIG. 5, there is shown a block diagram of a receiver that downconverts signals modulated using first, second, third and fourth modulation formats to a common intermediate frequency range using two downconverters, in accordance with an alternative preferred embodiment of the present invention. The receiver shown in FIG. 5 is substantially the same as the receiver of FIG. 4, except in the receiver of FIG. 5 only two channel selection filters 60, 62A are used to process the downconverted signals output by the two downconverters. Thus, in the receiver of FIG. 5, the second channel selection filter 62A filters downconverted signals modulated in accordance with the second modulation format (e.g., CDMA signals) output by downconverter 20, the second channel filter 62A also filters downconverted signals modulated in accordance with the third modulation format (e.g., PCS signals) output by the downconverter 24, and the second channel selection filter 62A further filters downconverted signals modulated in accordance with the fourth modulation format (e.g., GPS signals) output by downconverter 24.

Referring now to FIG. 6, there is shown a block diagram of a receiver that downconverts signals modulated using first, second, third and

fourth modulation formats to a common intermediate frequency range using two downconverters, in accordance with a further alternative preferred embodiment of the present invention. The receiver shown in FIG. 6 is substantially the same as the receiver of FIG. 4, except in the receiver of FIG. 6
5 only a single channel selection filter 60A is used to process the downconverted signals output by the two downconverters.

The previous description of the preferred embodiments is provided to enable any person skilled in the art to make or use the present invention. Although the present invention has been described in conjunction
10 with a mobile radio telephone receiver, the principles of the present invention may be applied in other contexts and applications. In addition, various modifications to the embodiments described above will be readily apparent to those skilled in the art, and the generic principles defined herein may be applied to other embodiments without the use of inventive faculty. Thus, the
15 present invention is not intended to be limited to the methods and apparatuses shown herein but is to be accorded the widest scope consistent with the claims set forth below.

What is claimed is:

CLAIMS

1. A receiver that downconverts signals modulated using a plurality of modulation formats to a common intermediate frequency range, comprising:
- a first band selection filter that selects a first frequency range;
 - a second band selection filter that selects a second frequency range;
 - a third band selection filter that selects a third frequency range;
 - a first downconverter, coupled to an output of the first band selection filter, that downconverts signals from the first frequency range to the common intermediate frequency range;
 - a second downconverter, selectively coupled by a switch to either an output of the second band selection filter or an output of the third band selection filter, that downconverts signals from either the second frequency range or the third frequency range to the common intermediate frequency range, wherein the second downconverter has an input coupled to a frequency doubling circuit;
 - switching circuitry that selectively couples one of either a first oscillating signal from a voltage controller oscillator (VCO) having a VCO frequency range or a second oscillating signal at a second frequency that is outside the VCO frequency range to an input of the first downconverter and an input of the frequency doubling circuit.
2. The receiver of claim 1, wherein the plurality of modulation formats include first, second, third and fourth modulation formats, the first and second modulation formats are transmitted to the receiver in the first frequency range, the third modulation format is transmitted to the receiver in the second frequency range, and the fourth modulation format is transmitted to the receiver in the third frequency range.

3. The receiver of claim 2, wherein the VCO is responsive to
2 control circuitry and selectively generates the first oscillating signal at a channel
frequency associated with the first modulation format, and the first
4 downconverter mixes the output of the first band selection filter with the first
oscillating signal at the channel frequency associated with the first modulation
6 format in order to downconvert signals modulated in accordance with the first
modulation format from the first frequency range to the common intermediate
8 frequency range.

4. The receiver of claim 3, wherein the VCO further
2 selectively generates the first oscillating signal at a channel frequency
associated with the second modulation format, and the first downconverter
4 mixes the output of the first band selection filter with the first oscillating signal
at the channel frequency associated with the second modulation format in order
6 to downconvert signals modulated in accordance with the second modulation
format from the first frequency range to the common intermediate frequency
8 range.

5. The receiver of claim 4, wherein the VCO further
2 selectively generates the first oscillating signal at a channel frequency
associated with the third modulation format, wherein the first oscillating signal
4 at the channel frequency associated with the third modulation format is
provided to the frequency doubling circuit, and the second downconverter
6 mixes the output of the second band selection filter with an output of the
frequency doubling circuit when the first oscillating signal at a channel
8 frequency associated with the third modulation format is provided to the
frequency doubling circuit in order to downconvert signals modulated in
10 accordance with the third modulation format from the second frequency range
to the common intermediate frequency range.

6. The receiver of claim 5, further comprising a mixer that
2 forms the second oscillating signal at the second frequency by offsetting the
frequency of the first oscillating signal.

7. The receiver of claim 6, wherein the second oscillating
2 signal is selectively provided to the frequency doubling circuit, and the second
downconverter mixes the output of the third band selection filter with the
4 output of the frequency doubling circuit when the second oscillating signal is
provided to the frequency doubling circuit in order to downconvert signals
6 modulated in accordance with the fourth modulation format from the third
frequency range to the common intermediate frequency range.

8. The receiver of claim 7, wherein the first modulation
2 format corresponds to analog modulation, the second modulation format
corresponds to CDMA modulation, the third modulation format corresponds to
4 PCS modulation, the fourth modulation format corresponds to GPS
modulation, the first frequency range is 869-894 MHz, the second frequency
6 range is 1930-1990 MHz, and the third frequency range is 1574.42-1576.42 MHz.

9. The receiver of claim 8, wherein the common intermediate
2 frequency range is centered about 183.48 MHz.

10. The receiver of claim 8, wherein the first oscillating signal
2 at a channel frequency associated with the first modulation format has a
frequency range of 1052.52 – 1077.45 MHz.

11. The receiver of claim 10, wherein the first oscillating signal
2 at a channel frequency associated with the second modulation format has a
frequency range of 1052.52 – 1077.45 MHz.

12. The receiver of claim 11, wherein the first oscillating signal
2 at a channel frequency associated with the third modulation format has a
frequency range of 1056.74 – 1086.74 MHz.

13. The receiver of claim 12, wherein the second oscillating
2 signal at a channel frequency associated with the fourth modulation format has
a frequency range centered at 879.45 MHz.

14. The receiver of claim 6, further comprising:
2 one or more channel selection filters, coupled to outputs of the
first and second downconverters, that filter the downconverted signals output
4 by the first and second downconverters;

a sampling circuit, coupled to an output of the one or more
6 channel selection filters, that selectively samples the downconverted signals
with a sampling clock that alternatively samples signals at either a first or
8 second sampling rate; and

wherein the sampling clock is coupled to a third oscillating signal
10 at a third frequency, and the third oscillating signal is an input to the mixer that
forms the second oscillating signal at the second frequency.

15. The receiver of claim 14, wherein the first sampling rate
2 provided by the sampling clock is equal to the third frequency divided by x ,
and the second sampling rate provided by the sampling clock is equal to the
4 third frequency divided by y , wherein x and y are integers.

16. The receiver of claim 15, wherein the first sampling rate
2 provided by the sampling clock is $1/3$ the third frequency, and the second
sampling rate provided by the sampling clock is one-fifteenth of the third
4 frequency.

17. The receiver of claim 15, wherein the sampling circuit
2 samples downconverted signals modulated in accordance with either the
second, third or fourth modulation formats in accordance with the first
4 sampling rate, and the sampling circuit samples downconverted signals
modulated in accordance with the first modulation format in accordance with
6 the second sampling rate.

18. The receiver of claim 17, wherein the one or more channel
2 selection filters include only a single channel selection filter.

19. The receiver of claim 17, wherein the one or more channel
2 selection filters include first and second channel selection filters, the first
channel selection filter filters downconverted signals modulated in accordance
4 with the first modulation format output by the first downconverter, the second
channel selection filter filters downconverted signals modulated in accordance
6 with the second modulation format output by the first downconverter, the
second channel filter also filters downconverted signals modulated in
8 accordance with the third modulation format output by the second
downconverter, and the second channel selection filter further filters
10 downconverted signals modulated in accordance with the fourth modulation
format output by the second downconverter.

20. The receiver of claim 17, wherein the one or more channel
2 selection filters include first, second and third channel selection filters, the first
channel selection filter filters downconverted signals modulated in accordance
4 with the first modulation format output by the first downconverter, the second
channel selection filter filters downconverted signals modulated in accordance
6 with the second modulation format output by the first downconverter, the
second channel filter also filters downconverted signals modulated in
8 accordance with the third modulation format output by the second
downconverter, and the third channel selection filter filters downconverted
10 signals modulated in accordance with the fourth modulation format output by
the second downconverter.

21. The receiver of claim 14, further comprising a plurality of
2 anti-alias filters coupled to the output of the one or more channel selection
filters and to an input of the sampling circuit.

22. The receiver of claim 1, further comprising:
2 a third downconverter, coupled to an output of the first band
selection filter, that downconverts signals from the first frequency range to the
4 common intermediate frequency range;
wherein the switching circuitry selectively couples one of either
6 the first oscillating signal from the VCO or the second oscillating signal to
inputs of the first and third downconverters and an input of the frequency
8 doubling circuit.

23. The method of claim 22, wherein the plurality of
2 modulation formats include first, second, third and fourth modulation formats,
the first and second modulation formats are transmitted to the receiver in the
4 first frequency range, the third modulation format is transmitted to the receiver
in the second frequency range, and the fourth modulation format is transmitted
6 to the receiver in the third frequency range.

24. The receiver of claim 23, wherein the VCO is responsive to
2 control circuitry and selectively generates the first oscillating signal at a channel
frequency associated with the first modulation format, and the first
4 downconverter mixes the output of the first band selection filter with the first
oscillating signal at the channel frequency associated with the first modulation
6 format in order to downconvert signals modulated in accordance with the first
modulation format from the first frequency range to the common intermediate
8 frequency range.

25. The receiver of claim 24, wherein the VCO further
2 selectively generates the first oscillating signal at a channel frequency
associated with the second modulation format, and the third downconverter
4 mixes the output of the first band selection filter with the first oscillating signal
at the channel frequency associated with the second modulation format in order
6 to downconvert signals modulated in accordance with the second modulation
format from the first frequency range to the common intermediate frequency
8 range.

26. The receiver of claim 25, wherein the VCO further
2 selectively generates the first oscillating signal at a channel frequency
associated with the third modulation format, wherein the first oscillating signal
4 at the channel frequency associated with the third modulation format is
provided to the frequency doubling circuit, and the second downconverter
6 mixes the output of the second band selection filter with an output of the
frequency doubling circuit when the first oscillating signal at a channel
8 frequency associated with the third modulation format is provided to the
frequency doubling circuit in order to downconvert signals modulated in
10 accordance with the third modulation format from the second frequency range
to the common intermediate frequency range.

27. The receiver of claim 26, further comprising a mixer that
2 forms the second oscillating signal at the second frequency by offsetting the
frequency of the first oscillating signal.

28. The receiver of claim 27, wherein the second oscillating
2 signal is selectively provided to the frequency doubling circuit, and the second
downconverter mixes the output of the third band selection filter with the
4 output of the frequency doubling circuit when the second oscillating signal is
provided to the frequency doubling circuit in order to downconvert signals
6 modulated in accordance with the fourth modulation format from the third
frequency range to the common intermediate frequency range.

29. The receiver of claim 28, wherein the first modulation
2 format corresponds to analog modulation, the second modulation format
corresponds to CDMA modulation, the third modulation format corresponds to
4 PCS modulation, the fourth modulation format corresponds to GPS
modulation, the first frequency range is 869-894 MHz, the second frequency
6 range is 1930-1990 MHz, and the third frequency range is 1574.42-1576.42 MHz.

30. The receiver of claim 29, wherein the common intermediate
2 frequency range is centered about 183.48 MHz.

31. The receiver of claim 29, wherein the first oscillating signal
2 at a channel frequency associated with the first modulation format has a
frequency range of 1052.52 – 1077.45 MHz.

32. The receiver of claim 31, wherein the first oscillating signal
2 at a channel frequency associated with the second modulation format has a
frequency range of 1052.52 – 1077.45 MHz.

33. The receiver of claim 32, wherein the first oscillating signal
2 at a channel frequency associated with the third modulation format has a
frequency range of 1056.74 – 1086.74 MHz.

34. The receiver of claim 33, wherein the second oscillating
2 signal at a channel frequency associated with the fourth modulation format has
a frequency range centered at 879.45 MHz.

35. The receiver of claim 27, further comprising:
2 one or more channel selection filters, coupled to outputs of the
first, second and third downconverters, that filter the downconverted signals
4 output by the first, second and third downconverters;
a sampling circuit, coupled to an output of the one or more
6 channel selection filters, that selectively samples the downconverted signals
with a sampling clock that alternatively samples signals at either a first or
8 second sampling rate; and
wherein the sampling clock is coupled to a third oscillating signal
10 at a third frequency, and the third oscillating signal is an input to the mixer that
forms the second oscillating signal at the second frequency.

36. The receiver of claim 35, wherein the first sampling rate
2 provided by the sampling clock is equal to the third frequency divided by x,
and the second sampling rate provided by the sampling clock is equal to the
4 third frequency divided by y, wherein x and y are integers.

37. The receiver of claim 36, wherein the first sampling rate
2 provided by the sampling clock is 1/3 the third frequency, and the second

4 sampling rate provided by the sampling clock is one-fifteenth of the third frequency.

2 38. The receiver of claim 36, wherein the sampling circuit
4 samples downconverted signals modulated in accordance with either the
6 second, third or fourth modulation formats in accordance with the first
sampling rate, and the sampling circuit samples downconverted signals
modulated in accordance with either the first modulation format in accordance
with the second sampling rate.

2 39. The receiver of claim 38, wherein the one or more channel
selection filters include only a single channel selection filter.

2 40. The receiver of claim 38, wherein the one or more channel
4 selection filters include first and second channel selection filters, the first
6 channel selection filter filters downconverted signals modulated in accordance
8 with the first modulation format output by the first downconverter, the second
10 channel selection filter filters downconverted signals modulated in accordance
with the second modulation format output by the third downconverter, the
second channel filter also filters downconverted signals modulated in
accordance with the third modulation format output by the second
downconverter, and the second channel selection filter further filters
downconverted signals modulated in accordance with the fourth modulation
format output by the second downconverter.

2 41. The receiver of claim 38, wherein the one or more channel
4 selection filters include first, second and third channel selection filters, the first
6 channel selection filter filters downconverted signals modulated in accordance
8 with the first modulation format output by the first downconverter, the second
channel selection filter filters downconverted signals modulated in accordance
with the second modulation format output by the third downconverter, the
second channel filter also filters downconverted signals modulated in
accordance with the third modulation format output by the second
downconverter, and the third channel selection filter filters downconverted

10 signals modulated in accordance with the fourth modulation format output by
the second downconverter.

42. The receiver of claim 35, further comprising a plurality of
2 anti-alias filters coupled to the output of the one or more channel selection
filters and to an input of the sampling circuit.

43. A method for downconverting signals modulated using a
2 plurality of modulation formats to a common intermediate frequency range,
comprising the steps of:

4 applying the input signals to a first band selection filter that
selects a first frequency range;

6 applying the input signals to a second band selection filter that
selects a second frequency range;

8 applying the input signals to a third band selection filter that
selects a third frequency range;

10 selectively providing one of either a first oscillating signal from a
voltage controller oscillator (VCO) having a VCO frequency range or a second
12 oscillating signal at a second frequency that is outside the VCO frequency range
to an input of a first downconverter and an input of a frequency doubling
14 circuit;

downconverting, with the first downconverter, signals from the
16 first frequency range to the common intermediate frequency range;

downconverting, with a second downconverter, signals from
18 either the second frequency range or the third frequency range to the common
intermediate frequency range, wherein the second downconverter has an input
20 coupled to the frequency doubling circuit, and the second downconverter is
selectively coupled by a switch to either an output of the second band selection
22 filter or an output of the third band selection filter.

44. An apparatus for downconverting signals modulated using
2 a plurality of modulation formats to a common intermediate frequency range,
comprising:

4 means for applying the input signals to a first band selection filter
that selects a first frequency range;

6 means for applying the input signals to a second band selection
filter that selects a second frequency range;

8 means for applying the input signals to a third band selection
filter that selects a third frequency range;

10 means for selectively providing one of either a first oscillating
signal from a voltage controller oscillator (VCO) having a VCO frequency range
12 or a second oscillating signal at a second frequency that is outside the VCO
frequency range to an input of a first downconverter and an input of a
14 frequency doubling circuit;

means for downconverting, with the first downconverter, signals
16 from the first frequency range to the common intermediate frequency range;

means for downconverting, with a second downconverter, signals
18 from either the second frequency range or the third frequency range to the
common intermediate frequency range, wherein the second downconverter has
20 an input coupled to the frequency doubling circuit, and the second
downconverter is selectively coupled by a switch to either an output of the
22 second band selection filter or an output of the third band selection filter.

45. A receiver that downconverts signals modulated using a
2 plurality of modulation formats to an intermediate frequency range,
comprising:

4 at least one downconverter that downconverts signals modulated
using the plurality of modulation formats to the intermediate frequency range,
6 wherein the intermediate frequency range is centered about an intermediate
frequency (f_{IF}); and

8 a sampling circuit, coupled to the at least one downconverter, that
samples the downconverted signals with a sampling clock that operates at a
10 sampling rate (F_s);

wherein $f_{IF} = [(2k + 1)/4] * (F_s)$, and k is an integer that is greater
12 or equal to 0.

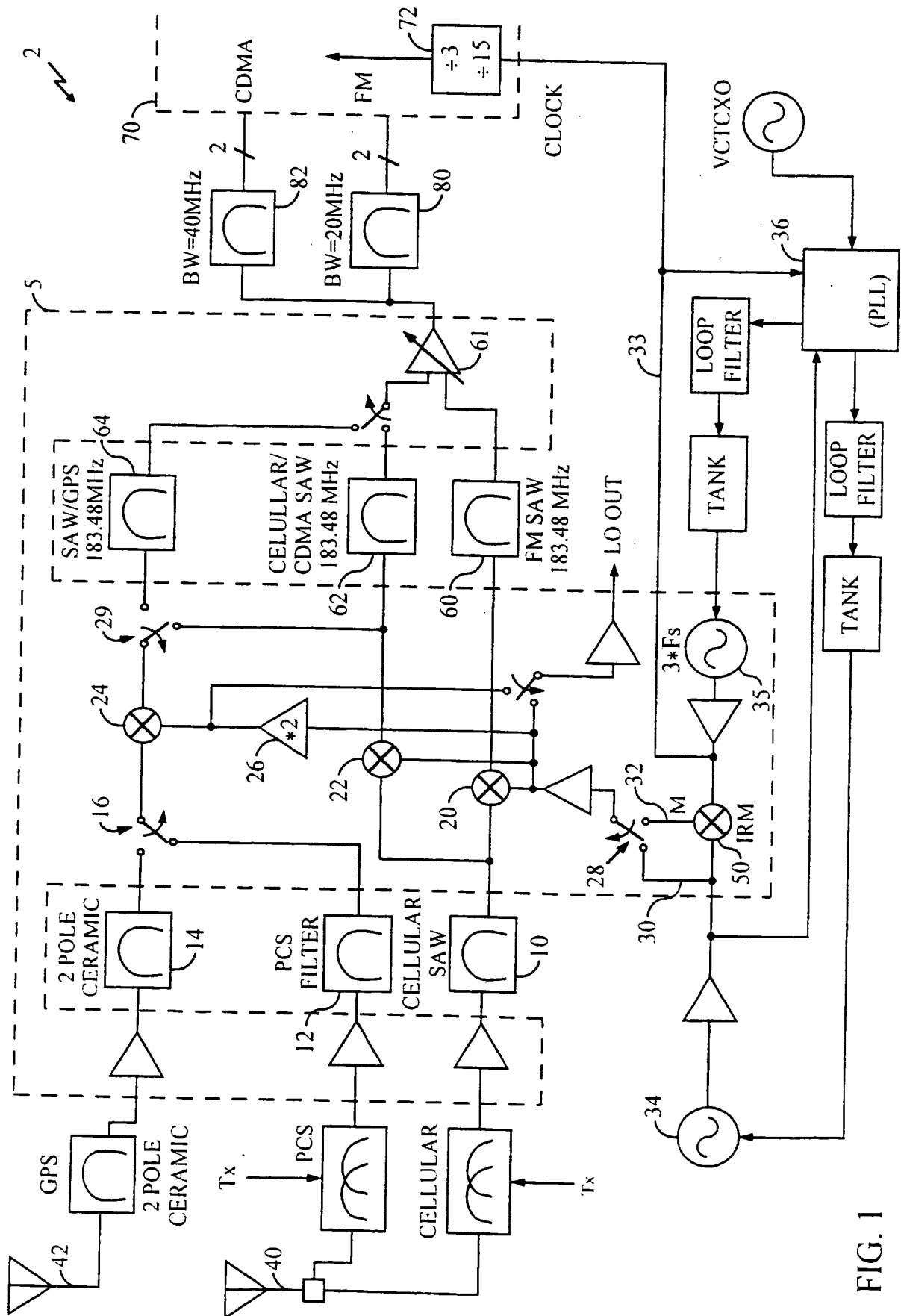


FIG. 1

2/6

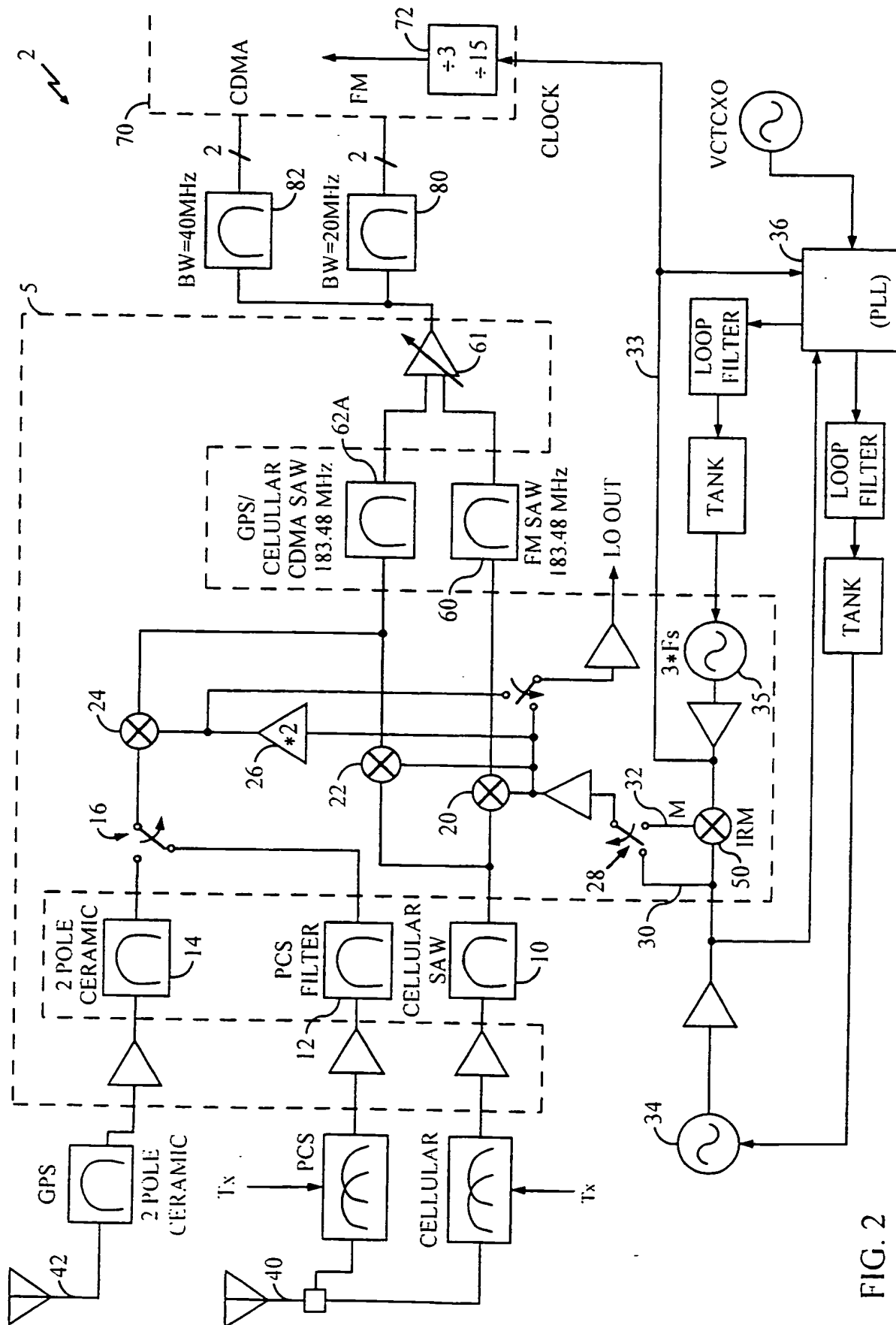


FIG. 2

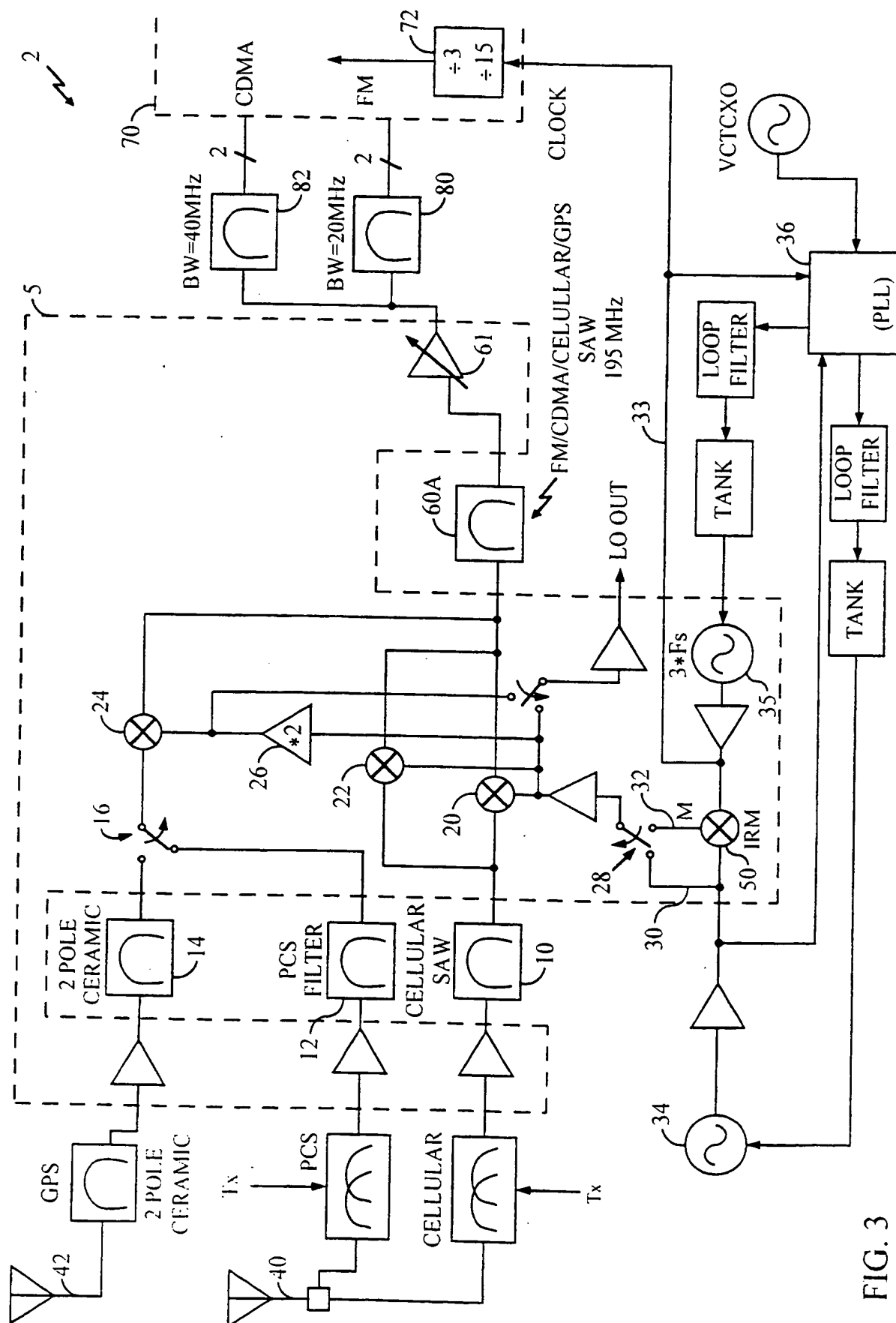


FIG. 3

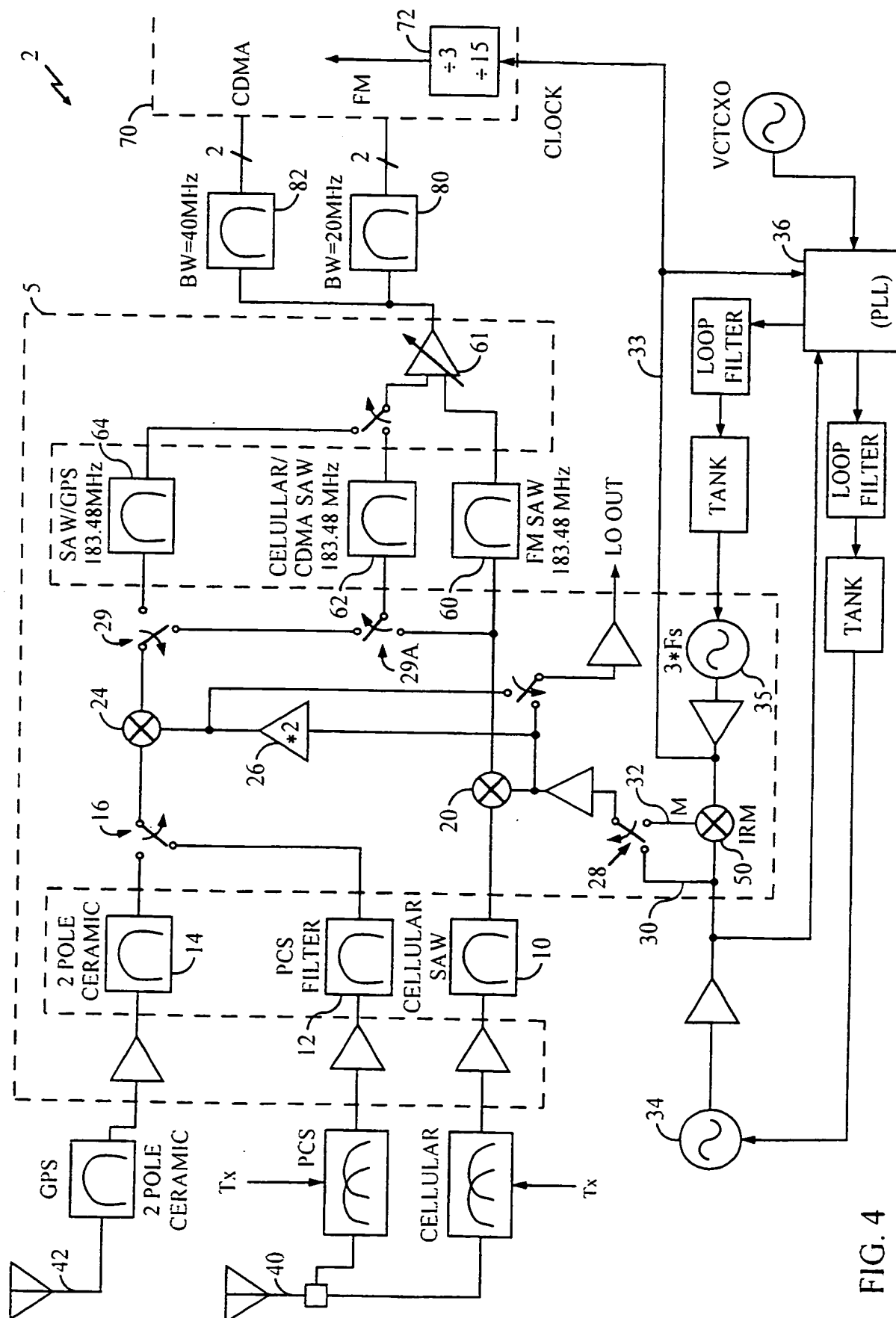


FIG. 4

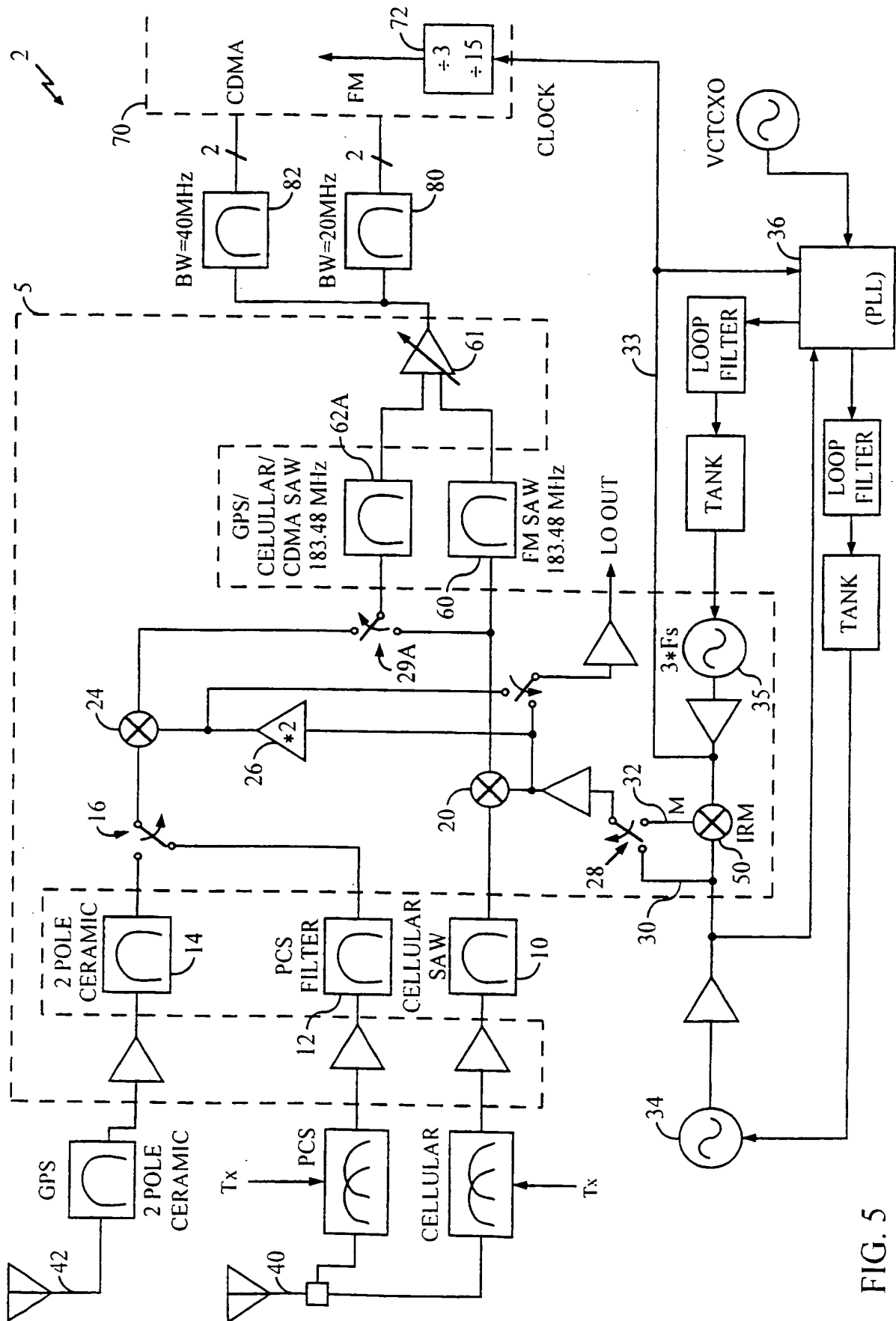


FIG. 5

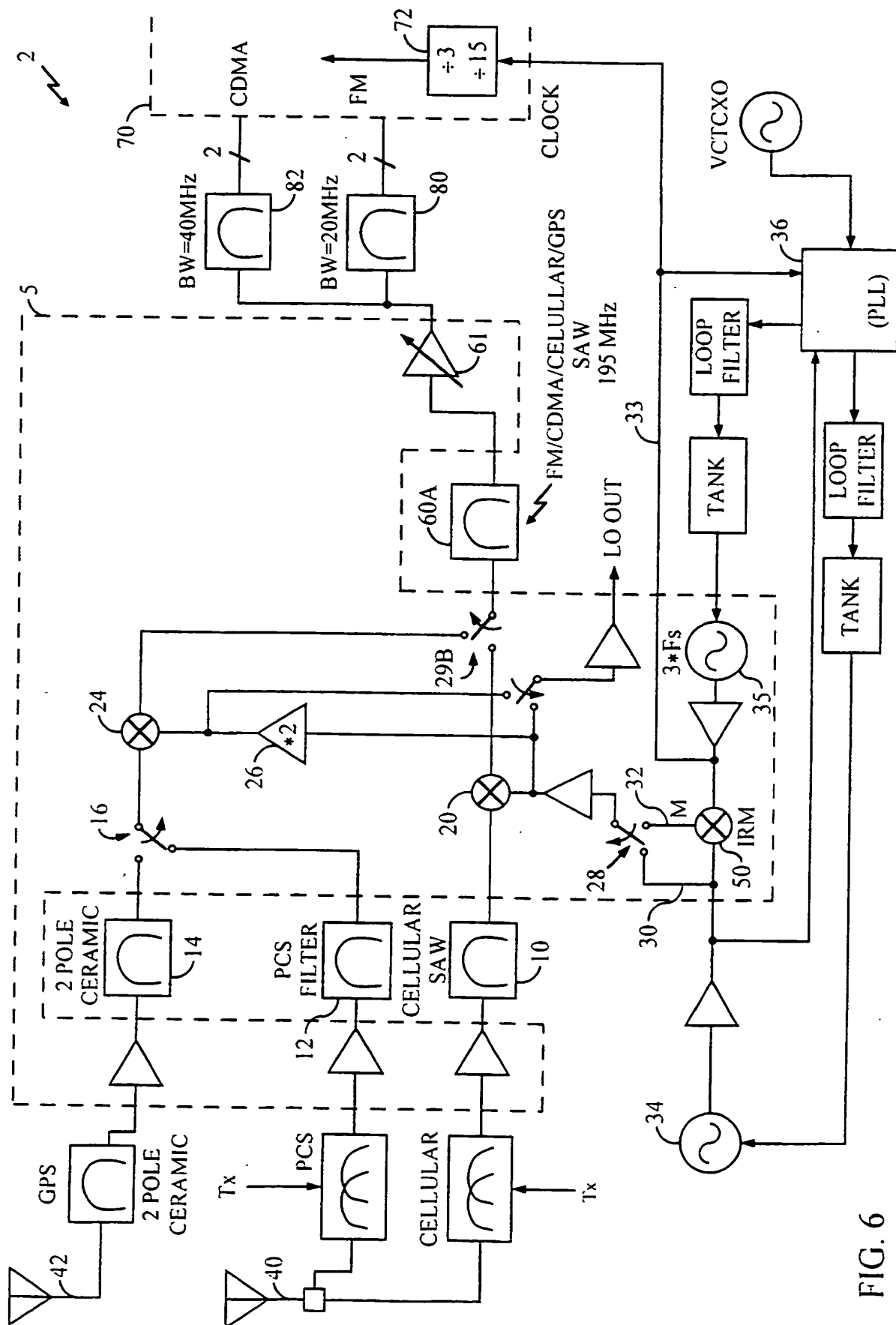


FIG. 6

INTERNATIONAL SEARCH REPORT

International Application No

PCT/US 99/04545

A. CLASSIFICATION OF SUBJECT MATTER
IPC 6 H0481/40

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 6 H048

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	WO 98 00927 A (ERICSSON GE MOBILE INC) 8 January 1998 see abstract see page 15, line 13 - page 20, line 8 see figure 6 ---	1-13, 43-45
A	WO 97 30523 A (NORTHERN TELECOM LTD ; BROWN DAVID ALAN (GB); NISBET JOHN JACKSON () 21 August 1997 see abstract see page 17, line 4 - page 21, line 18 see figure 6 see figure 7 --- -/--	1-5, 43-45



Further documents are listed in the continuation of box C.



Patent family members are listed in annex.

* Special categories of cited documents :

"A" document defining the general state of the art which is not considered to be of particular relevance

"E" earlier document but published on or after the international filing date

"L" document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)

"O" document referring to an oral disclosure, use, exhibition or other means

"P" document published prior to the international filing date but later than the priority date claimed

"T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention

"X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

"Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.

"&" document member of the same patent family

Date of the actual completion of the international search

5 July 1999

Date of mailing of the international search report

14/07/1999

Name and mailing address of the ISA

European Patent Office, P.B. 5818 Patentlaan 2
NL - 2280 HV Rijswijk
Tel. (+31-70) 340-2040, Tx. 31 651 epo nl.
Fax: (+31-70) 340-3016

Authorized officer

Lindhardt, U

INTERNATIONAL SEARCH REPORT

Int'l Application No

PCT/US 99/04545

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	EP 0 823 790 A (NOKIA MOBILE PHONES LTD) 11 February 1998 see abstract see column 5, line 51 - column 7, line 52 see figure 3 ---	1-5, 22, 43-45
A	EP 0 793 356 A (NOKIA MOBILE PHONES LTD) 3 September 1997 see abstract see column 2, line 11 - column 5, line 53 see figure 1 see figure 2 -----	1-5, 43-45

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PCT/US 99/04545

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
WO 9800927 A	08-01-1998	US 5732330 A AU 3411797 A EP 0909485 A	24-03-1998 21-01-1998 21-04-1999
WO 9730523 A	21-08-1997	GB 2310342 A EP 0880830 A	20-08-1997 02-12-1998
EP 0823790 A	11-02-1998	US 5794159 A JP 10107678 A	11-08-1998 24-04-1998
EP 0793356 A	03-09-1997	FI 960947 A US 5852603 A	30-08-1997 22-12-1998